

Parametric Sparse Sampling: Theory and Applications

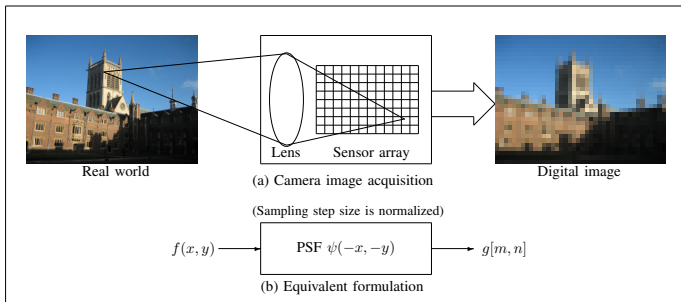
Pier Luigi Dragotti

June 19, 2014¹

¹This research is supported by European Research Council ERC, project 277800 (RecoSamp)



Sparsity and Sampling: Is This Relevant?



- ▶ The lens blurs the image.
- ▶ The image is sampled ('pixelized') by the CCD array.
- ▶ You want sharper and higher resolution images given the available pixels



Motivation: Image Resolution Enhancement



pixels



interpolation

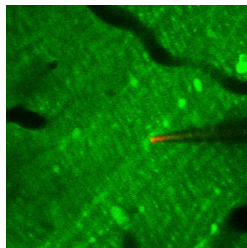
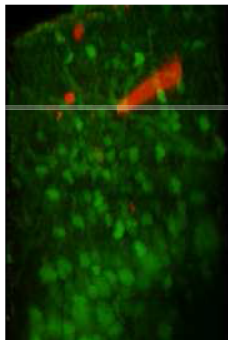


enhancement with sparsity priors



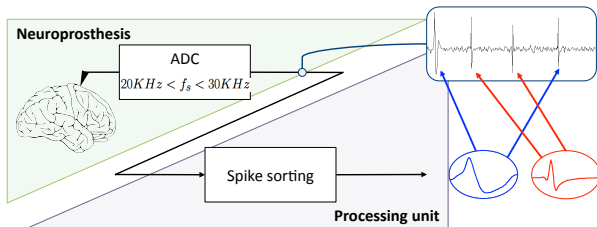
Motivation: Application in Neuroscience

Time resolution enhancement and calcium transient detection in multi-photon calcium imaging.



Motivation: Brain Machine Interface

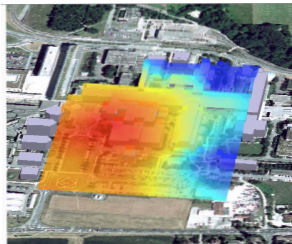
Applications in Neuroscience: Spike Sorting at sub-Nyquist rates



Wireless brain-machine interface place extreme limits on sampling.



Motivation: Sensor Networks



- ▶ Can we localise diffusion sources and estimate their activation time using sensor networks?
- ▶ Application:
 1. Check whether your government is lying ;-)
 2. Monitor dispersion in factories producing bio-chemicals



Problem Statement

What do all these problems have in common?

- ▶ The source is normally continuous in time and/or space (discretising it might not be an effective strategy)
- ▶ Measurements are discrete (e.g., pixels in a camera, sensors measurements)
- ▶ The observation process involves deterministic smoothing functions normally known a priori (e.g., point spread function in a camera, the diffusion kernel for diffusion fields)



Problem Statement

What do all these problems have in common?

- ▶ The source is normally continuous in time and/or space (discretising it might not be an effective strategy)
- ▶ Measurements are discrete (e.g., pixels in a camera, sensors measurements)
- ▶ The observation process involves deterministic smoothing functions normally known a priori (e.g., point spread function in a camera, the diffusion kernel for diffusion fields)

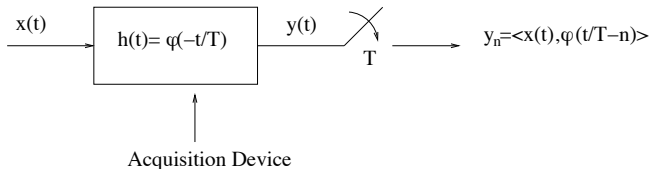
Our Approach

- ▶ From the samples, using the knowledge of the observation process, estimate proper integral measurements of the source (e.g., estimate the Fourier transform at specific frequencies)
- ▶ Given the integral measurements (e.g., partial Fourier transform), solve the inverse problem using sparsity priors



Problem Statement

You are given a class of functions. You have a sampling device. Given the measurements $y_n = \langle x(t), \varphi(t/T - n) \rangle$, you want to reconstruct $x(t)$.



Natural questions:

- ▶ When is there a one-to-one mapping between $x(t)$ and y_n ?
- ▶ What signals can be sampled and what kernels $\varphi(t)$ can be used?
- ▶ What reconstruction algorithm?

Signals with Finite Rate of Innovation

Consider a signal of the form:

$$x(t) = \sum_{k \in \mathbb{Z}} \gamma_k g(t - t_k). \quad (1)$$

- ▶ Given $g(t)$, the signal is completely specified by γ_k and t_k .
- ▶ **Key intuition:** if the number of samples is larger than the number of parameters then reconstruction is possible
- ▶ This is an '**analogue**' sparsity model



Signals with Finite Rate of Innovation

Consider a signal of the form:

$$x(t) = \sum_{k \in \mathbb{Z}} \gamma_k g(t - t_k). \quad (2)$$

The rate of innovation of $x(t)$ is then defined as

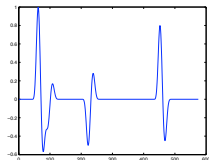
$$\rho = \lim_{\tau \rightarrow \infty} \frac{1}{\tau} C_x \left(-\frac{\tau}{2}, \frac{\tau}{2} \right), \quad (3)$$

where $C_x(-\tau/2, \tau/2)$ is a function counting the number of free parameters in the interval τ .

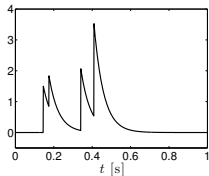
Definition [VetterliMB:02] A signal with a **finite rate of innovation** is a signal whose parametric representation is given in (2) and with a finite ρ as defined in (3).



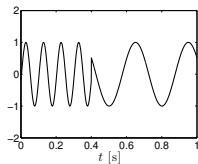
Examples of Signals with Finite Rate of Innovation



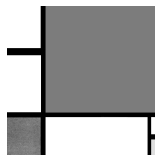
Filtered Streams of Diracs



Decaying Exponentials



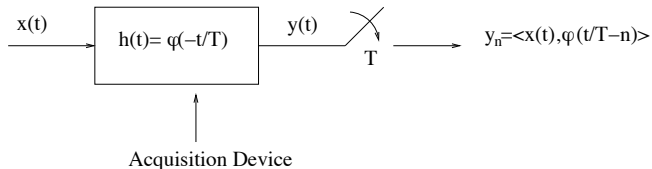
Piecewise Sinusoidal Signals



Mondrian paintings ;-)



Sampling Kernels



- ▶ Given by nature
 - ▶ Diffusion equation, Green function. Ex: sensor networks.
- ▶ Given by the set-up
 - ▶ Designed by somebody else. Ex: Hubble telescope, digital cameras.
- ▶ Given by design
 - ▶ Pick the best kernel. Ex: engineered systems.

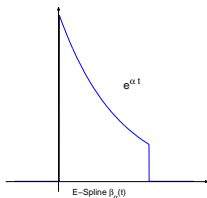


Sampling Kernels

Any kernel $\varphi(t)$ that can reproduce exponentials:

$$\sum_n c_{m,n} \varphi(t - n) = e^{\alpha_m t}, \quad \alpha_m = \alpha_0 + m\lambda \text{ and } m = 0, 1, \dots, L.$$

This includes any composite kernel of the form $\gamma(t) * \beta_{\vec{\alpha}}(t)$ where $\beta_{\vec{\alpha}}(t) = \beta_{\alpha_0}(t) * \beta_{\alpha_1}(t) * \dots * \beta_{\alpha_L}(t)$ and $\beta_{\alpha_i}(t)$ is an Exponential Spline of first order [UnserB:05].



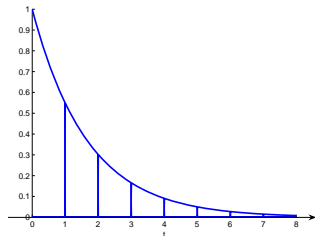
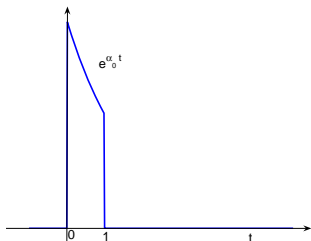
$$\beta_{\alpha}(t) \Leftrightarrow \hat{\beta}(\omega) = \frac{1 - e^{\alpha - j\omega}}{j\omega - \alpha}$$

Notice:

- ▶ α can be complex.
- ▶ E-Spline is of compact support.
- ▶ E-Spline reduces to the classical polynomial spline when $\alpha = 0$.



Exponential Reproducing Kernels



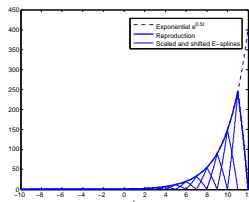
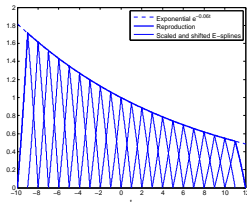
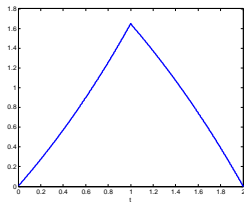
The E-spline of first order $\beta_{\alpha_0}(t)$ reproduces the exponential $e^{\alpha_0 t}$:

$$\sum_n c_{0,n} \beta_{\alpha_0}(t - n) = e^{\alpha_0 t}.$$

In this case $c_{0,n} = e^{\alpha_0 n}$. In general, $c_{m,n} = c_{m,0} e^{\alpha_m n}$.



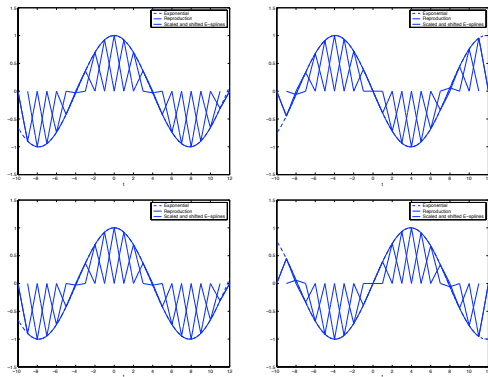
Exponential Reproducing Kernels



Here the E-spline is of second order and reproduces the exponential $e^{\alpha_0 t}$, $e^{\alpha_1 t}$: with $\alpha_0 = -0.06$ and $\alpha_1 = 0.5$.



Exponential Reproducing Kernels



Here $\vec{\alpha} = (-j\omega_0, j\omega_0)$ and $\omega_0 = 0.2$. $\sum_n c_{n,m} \beta_{\vec{\alpha}}(t-n) = e^{jm\omega_0 t}$ $m = -1, 1$.

Notice: $\beta_{\vec{\alpha}}(t)$ is a real function, but the coefficients $c_{m,n}$ are complex.



Generalised Strang-Fix Conditions

A function $\varphi(t)$ can reproduce the exponential:

$$e^{j\omega_m t} = \sum_n c_{m,n} \varphi(t - n)$$

if and only if

$$\hat{\varphi}(j\omega_m) \neq 0 \text{ and } \hat{\varphi}(j\omega_m + j2\pi l) = 0 \quad l \in \mathbb{Z} \setminus \{0\}$$

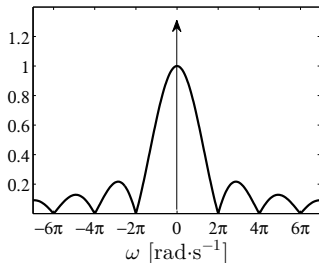
where $\hat{\varphi}(\cdot)$ is the Fourier transform of $\varphi(t)$.

Also note that $c_{m,n} = c_{m,0} e^{j\omega_m n}$ with $c_{m,0} = \hat{\varphi}(j\omega_m)^{-1}$.

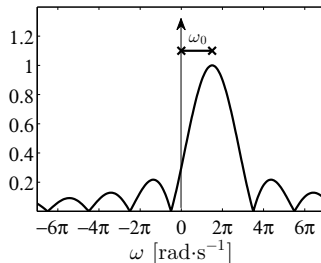


Generalised Strang-Fix Conditions

- ▶ Strang-Fix conditions are not restrictive
- ▶ Any low-pass filter approximately satisfies them.



(a) $|\hat{\beta}_\alpha(\omega)|$ with $\alpha = 0$

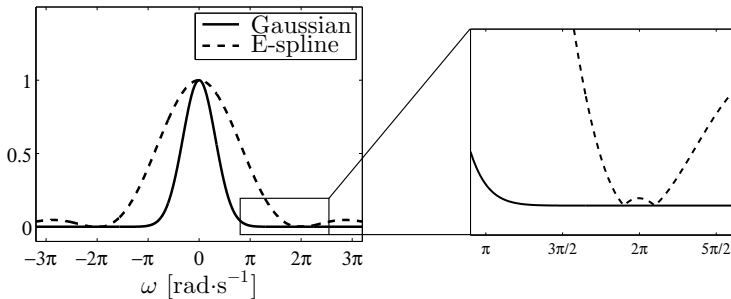


(b) $|\hat{\beta}_\alpha(\omega)|$ with $\alpha = i\omega_0$

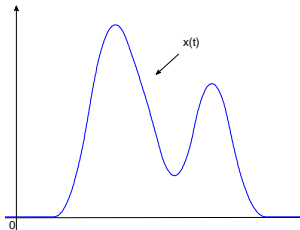


Approximate Strang-Fix

- ▶ Strang-Fix conditions are not restrictive
- ▶ Any low-pass filter approximately satisfies them.



From Samples to Integral Measurements



- ▶ Consider any $x(t)$ with $t \in [0, N)$ and sampling period $T = 1$.
- ▶ The sampling kernel $\varphi(t)$ satisfies

$$\sum_n c_{m,n} \varphi(t - n) = e^{j\omega_m t} \quad m = 1, \dots, L,$$

- ▶ We want to retrieve $x(t)$, from the samples $y_n = \langle x(t), \varphi(t - n) \rangle$, $n = 0, 1, \dots, N - 1$.



From Samples to Integral Measurements

We have that

$$\begin{aligned} s_m &= \sum_{n=0}^{N-1} c_{m,n} y_n \\ &= \langle x(t), \sum_{n=0}^{N-1} c_{m,n} \varphi(t-n) \rangle \\ &= \int_{-\infty}^{\infty} x(t) e^{j\omega_m t} dt, \quad m = 1, \dots, L. \end{aligned}$$

- ▶ Note that s_m is the Fourier transform of $x(t)$ evaluated at $j\omega_m$.



From Samples to Signals

- ▶ Consider FRI signals which are completely specified by a finite number of free parameters
- ▶ For classes of **parametrically** sparse signals there is a one-to-one mapping between samples and signal:

$$x(t) \Leftrightarrow \hat{x}(j\omega_m) \quad m = 1, 2, \dots, L$$

- ▶ The number d of degrees of freedom of the signal must satisfy $d \leq L$



Sampling Streams of Diracs

- ▶ Assume $x(t)$ is a stream of K Diracs on the interval of size N :
 $x(t) = \sum_{k=0}^{K-1} x_k \delta(t - t_k)$, $t_k \in [0, N)$.
- ▶ We restrict $j\omega_m = j\omega_0 + jm\lambda$ $m = 1, \dots, L$ and $L \geq 2K$.
- ▶ We have N samples: $y_n = \langle x(t), \varphi(t - n) \rangle$, $n = 0, 1, \dots, N-1$:
- ▶ We obtain

$$\begin{aligned} s_m &= \sum_{n=0}^{N-1} c_{m,n} y_n \\ &= \int_{-\infty}^{\infty} x(t) e^{j\omega_m t} dt, \\ &= \sum_{k=0}^{K-1} x_k e^{j\omega_m t_k} \\ &= \sum_{k=0}^{K-1} \hat{x}_k e^{j\lambda m t_k} = \sum_{k=0}^{K-1} \hat{x}_k u_k^m, \quad m = 1, \dots, L. \end{aligned}$$



Prony's Method

- ▶ The quantity

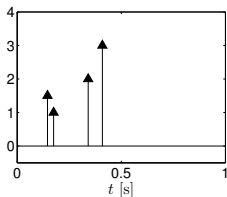
$$s_m = \sum_{k=0}^{K-1} \hat{x}_k u_k^m, \quad m = 1, \dots, L$$

is a sum of exponentials.

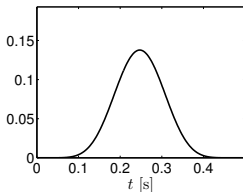
- ▶ Retrieving the locations u_k and the amplitudes \hat{x}_k from $\{s_m\}_{m=1}^L$ is a classical problem in spectral estimation and was first solved by Gaspard de Prony in 1795.
- ▶ Given the pairs $\{u_k, \hat{x}_k\}$, then $t_k = (\ln u_k)/\lambda$ and $x_k = \hat{x}_k/e^{\alpha_0 t_k}$.



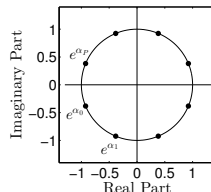
Sampling Streams of Diracs: Numerical Example



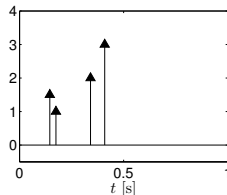
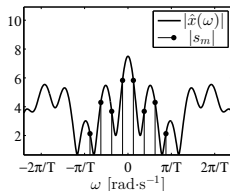
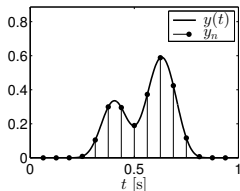
(a) Input signal, $x(t)$



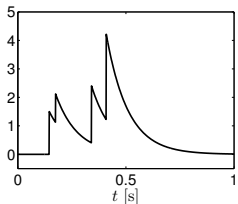
(b) Sampling kernel, $h(t)$



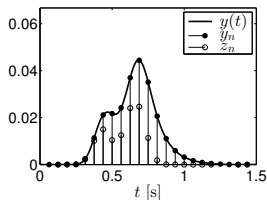
(c) $e^{\alpha t}$ reproduced by $h(t)$



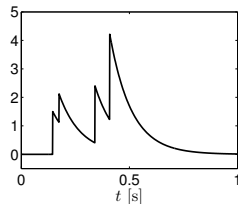
Stream of Decaying Exponentials



(a) Input signal, $x(t)$



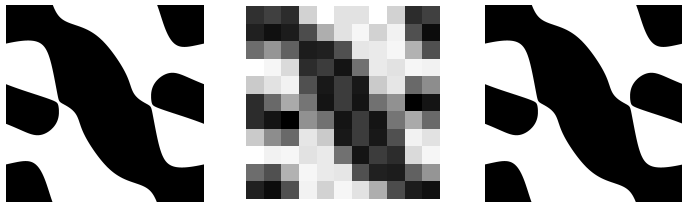
(b) Filtered and sampled signal



(c) Reconstructed signal



Sampling 2-D domains



The curve is implicitly defined through the equation [PanBluDragotti:11,14]:

$$f(x, y) = \sum_{k=1}^K \sum_{i=1}^I b_{k,i} e^{-j2\pi xk/M} e^{-j2\pi yi/N} = 0.$$

The coefficients $b_{k,i}$ are the only free parameters in the model.



Sampling 2-D domains



samples



interpolation



inter+ curve constraint

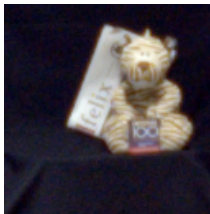


Application: Image Super-Resolution [BaboulazD:09]

Acquisition with Nikon D70



(a) Original (2014×3040)



(b) ROI (128×128)



(b) Super-res (1024×1024)



Application: Image Super-Resolution



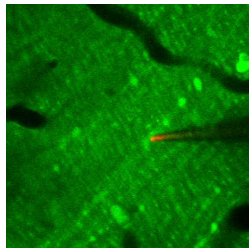
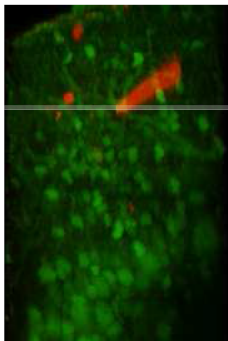
(a) Original (48×48)



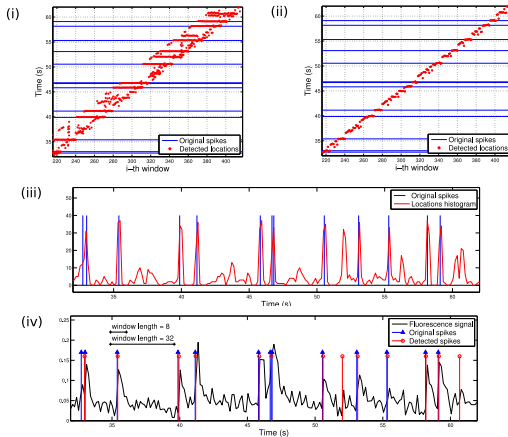
(b) Super-res (480×480)



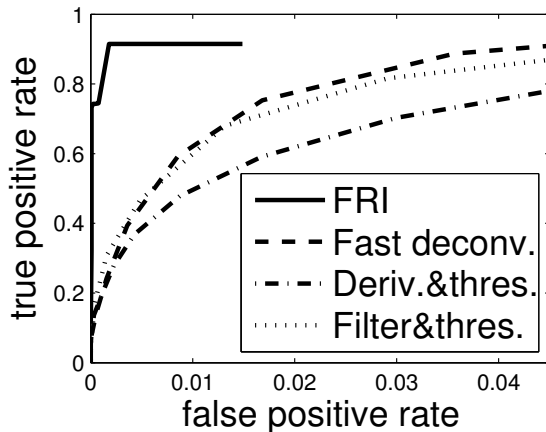
Neural Activity Detection [OnativiaSD:13]



Calcium Transient Detection



Calcium Transient Detection



Localisation of Diffusion Sources using Sensor Networks [Murray-BruceD:14]



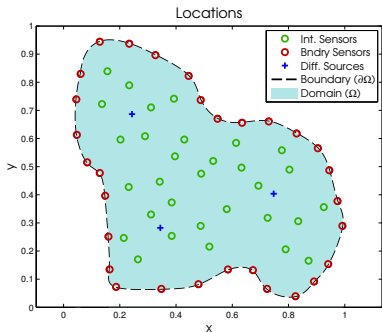
- ▶ The diffusion equation models the dispersion of chemical plumes, smoke from forest fires, radioactive materials
- ▶ The phenomenon is sampled in space and time using a sensor network.
- ▶ Sources often localised in space. Can we retrieve their location and the time of activation?



Localisation of Diffusion Sources using Sensor Networks

Good news:

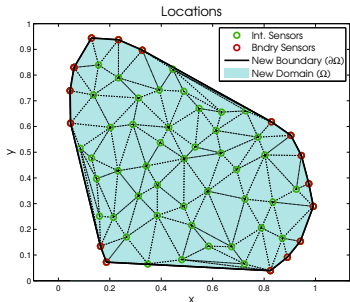
- ▶ When sources are localised in space and time, the field inversion is equivalent to a sparse sampling problem
- ▶ Proper linear combinations of sensors measurements in time and space leads to a Prony-type problem



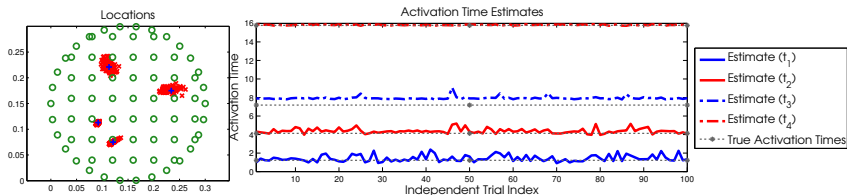
Localisation of Diffusion Sources using Sensor Networks

Good news:

- ▶ When sources are localised in space and time, the field inversion is equivalent to a sparse sampling problem
- ▶ Proper linear combinations of sensors measurements in time and space leads to a Prony-type problem



Localisation of Diffusion Sources: Numerical Results



(b) 100 independent trials using noisy sensor measurement samples (SNR=15dB).



Conclusions

Sampling signals using sparsity models:

- ▶ New framework that allows the sampling and reconstruction of continuous-time non-bandlimited signals.
- ▶ Use the knowledge of the acquisition process to map discrete measurements to specific integral measurements
- ▶ Use sparsity priors to reconstruct the original signal

Outlook:

- ▶ Promising applications in neuroscience, sensor networks, super-resolution imaging
- ▶ No **golden hammer**. Same framework but you need to fit the right model and carve the right solution for your problem: continuous/discrete, fast/complex, redundant/ not-redundant

Still many open questions from theory to practice!



References

On sampling

- ▶ J. Uriguen, T. Blu, and P.L. Dragotti 'FRI Sampling with Arbitrary Kernels', IEEE Trans. on Signal Processing, November 2013
- ▶ T. Blu, P.L. Dragotti, M. Vetterli, P. Marziliano and L. Coulot 'Sparse Sampling of Signal Innovations: Theory, Algorithms and Performance Bounds,' IEEE Signal Processing Magazine, vol. 25(2), pp. 31-40, March 2008
- ▶ P.L. Dragotti, M. Vetterli and T. Blu, 'Sampling Moments and Reconstructing Signals of Finite Rate of Innovation: Shannon meets Strang-Fix', IEEE Trans. on Signal Processing, vol.55 (5), pp.1741-1757, May 2007.
- ▶ J. Berent and P.L. Dragotti, and T. Blu, 'Sampling Piecewise Sinusoidal Signals with Finite Rate of Innovation Methods,' IEEE Transactions on Signal Processing, Vol. 58(2), pp. 613-625, February 2010.
- ▶ J. Uriguen, P.L. Dragotti and T. Blu, 'On the Exponential Reproducing Kernels for Sampling Signals with Finite Rate of Innovation' in Proc. of Sampling Theory and Application Conference, Singapore, May 2011.
- ▶ H. Pan, T. Blu, and P.L. Dragotti, 'Sampling Curves with Finite Rate of Innovation' IEEE Trans. on Signal Processing, January 2014.



References (cont'd)

On Image Super-Resolution

- ▶ L. Baboulaz and P.L. Dragotti, 'Exact Feature Extraction using Finite Rate of Innovation Principles with an Application to Image Super-Resolution', IEEE Trans. on Image Processing, vol.18(2), pp. 281-298, February 2009.

On Calcium Transient Detection

- ▶ Jon Onativia, Simon R. Schultz, and Pier Luigi Dragotti, A Finite Rate of Innovation algorithm for fast and accurate spike detection from two-photon calcium imaging, Journal of Neural Engineering, August 2013 .

On Diffusion Fields and Sensor Networks

- ▶ John Murray-Bruce and Pier Luigi Dragotti, Spatio-Temporal Sampling and Reconstruction of Diffusion Fields induced by Point Sources, Proc. of IEEE Conf. ICASSP, Florence (It), May 2014 .



Overview of Prony's Method

Assume: $y_n = \sum_{k=0}^{K-1} \alpha_k u_k^n$ and consider the polynomial:

$$P(x) = \prod_{k=1}^K (x - u_k) = x^K + h_1 x^{K-1} + h_2 x^{K-2} + \dots + h_{K-1} x + h_K.$$

It is easy to verify that

$$y_{n+K} + h_1 y_{n+K-1} + h_2 y_{n+K-2} + \dots + h_K y_n = \sum_{1 \leq k \leq K} \alpha_k u_k^n P(u_k) = 0.$$

In matrix-vector form for indices n such that $\ell \leq n < \ell + K$, we get

$$\begin{bmatrix} y_{\ell+K} & y_{\ell+K-1} & \cdots & y_{\ell} \\ y_{\ell+K+1} & y_{\ell+K} & \cdots & y_{\ell+1} \\ \vdots & \ddots & \ddots & \vdots \\ y_{\ell+2K-2} & \ddots & \ddots & \vdots \\ y_{\ell+2K-1} & y_{\ell+2K-2} & \cdots & y_{\ell+K-1} \end{bmatrix} \begin{bmatrix} 1 \\ h_1 \\ h_2 \\ \vdots \\ h_K \end{bmatrix} = \mathbf{T}_{K,\ell} \mathbf{h} = \mathbf{0}$$



Overview of Prony's Method

The vector of polynomial coefficients $\mathbf{h} = [1, h_1, \dots, h_K]^T$ is in the null space of $\mathbf{T}_{K,\ell}$. Moreover, $\mathbf{T}_{K,\ell}$ has size $K \times (K + 1)$ and has full row rank when the u_k 's are distinct. Therefore \mathbf{h} is unique. □

Prony's method summary:

1. Given the input y_n , build the Toeplitz matrix $\mathbf{T}_{K,\ell}$ and solve for \mathbf{h} . This can be achieved by taking the SVD of $\mathbf{T}_{K,\ell}$.
2. Find the roots of $P(x) = 1 + \sum_{n=1}^K h_n x^{K-n}$. These roots are exactly the exponentials $\{u_k\}_{k=0}^{K-1}$.
3. Given the $\{u_k\}_{k=0}^{K-1}$, find the corresponding amplitudes $\{\alpha_k\}_{k=0}^{K-1}$ by solving K linear equations.

